

Chapter 8. A/D Conversion and Data Acquisition

A/D conversion (or analog-to-digital conversion) is to read analog values from, for example, temperature or voltage level, into the PIC chip in the form of digital value. PIC 16F877 has an internal built-in module for this A/D conversion. A/D conversion has many applications: reading from a temperature sensor and displaying in on a PC screen of an LCD display; reading from a pressure sensor for blood pressure and acquiring the pressure data into a text file in a PC; reading a current value through an electric wire and alerting a circuit protection device for an above-normal power consumption, etc. In this chapter, we thoroughly study the built-in A/D module and practice of A/D conversion coding with a few practical applications.

1. A/D Conversion Module

A/D conversion is well utilized for external analog signal reading such as voltage, current, temperature, pressure, distance, or even color information. 16F877 has a A/D module. In this chapter, we will study the details of A/D converter module and its application. As need arises, some explanation on general A/D Conversion is discussed occasionally.

There is an 8 channel A-to-D (or A/D) converter module inside a 16F877: AN7 – AN0. These pins are not as well organized as other I/O pins. The lower four channels, AN0 – AN3, are arranged in the pin nos. 2 – 5, and AN4 – AN7 arrange with the pin Nos. of 7 through 10.

The A/D module allows conversion of an analog input signal to a corresponding 10-bit digital number. The output of the sample and hold is the input into the converter, which generates the result via successive approximation.

The analog reference voltages (positive and negative supply) are software selectable to either the device's supply voltages (AVDD, AVss) or the voltage level on the AN3/VREF+ and AN2/VREF-pins.

There are three types of registers we have to well control for A/D conversion. They are: A/D Result Registers (ADRESH and ADRESL), A/D Control Register0 (ADCON0), and A/D Control Register1 (ADCON1). The ADCON0 register controls the operation of the A/D module. The ADCON1 register configures the functions of the port pins. ADRESH and ADRESL registers contain the 10-bit results. Since each register is 8-bit register, we see that only one of the registers would be fully filled while the other would be partially filled by the A/D conversion result. Which register we configure to be fully filled, and which one to be partially filled is controlled by 'result justification': left- or right- justified.

Let's examine ADCON0 register first for the A/D operation.

The first two bits are assigned to select the A/D conversion clock. For correct A/D conversion, as the electrical specification of 16F877 states, the minimum A/D conversion clock must be selected to ensure a minimum of 1.6 μ s. With 20MHz crystal oscillation, the pre-scaled clock of $F_{osc}/2$ would be 100 ns, while the pre-scaled clock of $F_{osc}/8$ would be 400ns. So either selection would violate the minimum conversion clock of 1.6 μ s. The $F_{osc}/32$ with 1.6 μ s would satisfy the minimum clock. However, the internal RC source has typical 6 μ s of clock pulse. So, in 20MHz oscillator, selection of RC is safer and may be the only safe option for the A/D clock.

ADCON0 Register (1Fh)

ADCS1	ADCS0	CHS2	CHS1	CHS0	GO/DONE	0	ADON
						Always 0	

ADCS1:ADCS0: A/D Conversion Clock Select bits

00 = $F_{osc}/2$

01 = $F_{osc}/8$

10 = $F_{osc}/32$

11 = FRC (clock derived from internal A/D RC oscillator)

CHS2:CHS0: Analog Channel Select bits

000 = channel 0, (AN0) 001 = channel 1, (AN1)

010 = channel 2, (AN2) 011 = channel 3, (AN3)

100 = channel 4, (AN4) 101 = channel 5, (AN5)

110 = channel 6, (AN6) 111 = channel 7, (AN7)

GO/DONE: A/D Conversion Status bit

1 = A/D conversion in progress (**Start A/D**)

0 = A/D conversion not in progress

ADON: A/D On bit

1 = A/D converter **On**

0 = A/D converter **Off**

The next three bits select which one channel we use to read analog signal from, external world. Similar channel selection is done in ADCON1 to determine which channels are for analog input and which are for digital I/Os. Anyway, for ADCON0, select one channel you want to read. If you have multiple analog signals, you still have to select one channel for reading and then select another for another reading, etc. CS2:CS0=(000) would select the AN0 for the analog signal reading channel. The second bit (GO/~DONE) indicates the A/D conversion status: 1 indicates the process is still going on and 0 for no process. By setting the bit, we can start the A/D process. This bit is automatically cleared, when a process is finished, there is no need to clear the bit in program code. The last bit ADON works as a switch to turn on/off the A/D module: setting would make the A/D module ready for a conversion process. However, the final say is reserved to the GO/~DONE bit for actually starting the conversion.

For the ADCON1 register, we use only five bits: ADFM and PCFG3:PCFG0.

ADFM is to decide how we store the 10-bit A/D conversion result to the two A/D result registers: ADRESH and ADRESL. When set, the "Right Justification" is selected which stores the 8 LSBs of the result are stored to ADRESL and the 2 MSBs of the results are stored to 2 LSB positions of ADRESH. On the other hand, with its bit cleared, the "Left Justification" is selected which stores the 8 MSBs of the result into ADRESH register and the 2 LSBs of the result to the 2 MSB positions of ADRESL register. See the diagram below for illustration.

ADCON1 Register (9Fh)

----	----	ADFM	----	PCFG3	PCFG2	PCFG1	PCFG0
------	------	------	------	-------	-------	-------	-------

Read as '0' Read as '0' Read as '0'
ADFM: A/D Result format select
 1 = Right justified
 0 = left justified

PCFG3:PCFG0: A/D Port Configuration Control bits

PCFG	AN7	AN6	AN5	AN4	AN3	AN2	AN1	AN0	V _{REF+}	V _{REF-}	C / R
0000	A	A	A	A	A	A	A	A	AV _{DD}	AV _{SS}	8 / 0
0001	A	A	A	A	V _{REF+}	A	A	A	AN3	AV _{SS}	7 / 1
0010	D	D	D	A	A	A	A	A	AV _{DD}	AV _{SS}	5 / 0
0011	D	D	D	A	V _{REF+}	A	A	A	AN3	AV _{SS}	4 / 1
0100	D	D	D	D	A	D	A	A	AV _{DD}	AV _{SS}	3 / 0
0101	D	D	D	D	V _{REF+}	D	A	A	AN3	AV _{SS}	2 / 1
011x	D	D	D	D	D	D	D	D	---	---	0 / 0
1000	A	A	A	A	V _{REF+}	V _{REF-}	A	A	AN3	AN2	6 / 2
1001	D	D	D	A	A	A	A	A	AV _{DD}	AV _{SS}	6 / 0
1010	D	D	D	A	V _{REF+}	A	A	A	AN3	AV _{SS}	5 / 1
1011	D	D	D	A	V _{REF+}	V _{REF-}	A	A	AN3	AN2	4 / 2
1100	D	D	D	A	V _{REF+}	V _{REF-}	A	A	AN3	AN2	3 / 2
1101	D	D	D	D	V _{REF+}	V _{REF-}	A	A	AN3	AN2	2 / 2
1110	D	D	D	D	D	D	D	A	AV _{DD}	AV _{SS}	1 / 0
1111	D	D	D	D	V _{REF+}	V _{REF-}	D	A	AN3	AN2	1 / 2

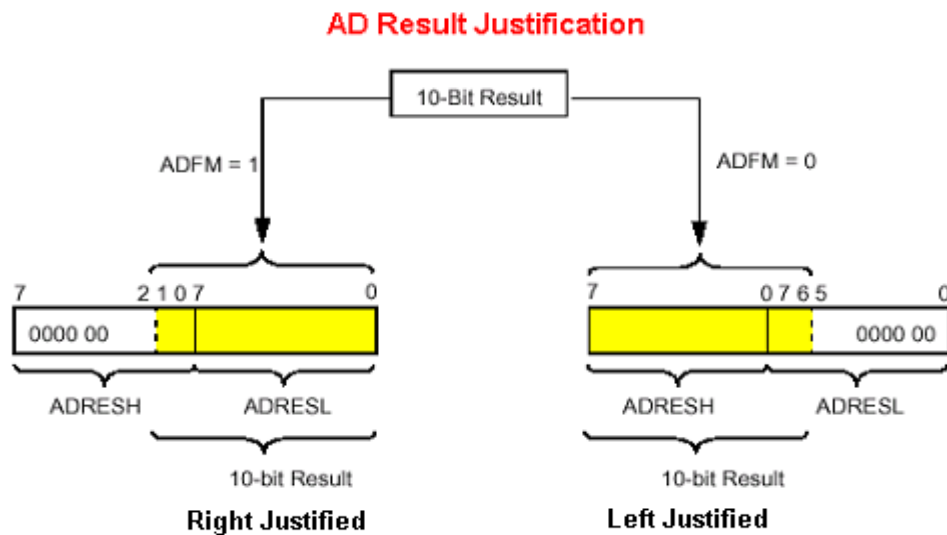


Fig. 58 AD Result Justification

By the way, how much difference do the different justifications make? For example, with left justification, let's assume that we ignore the ADRESL and use only the content of ADRESH. In other words, by this, we ignore the 2 LSBs of the result, and get only the 8 MSBs. Since the last two bits are not used, the resolution would be reduced by 4.

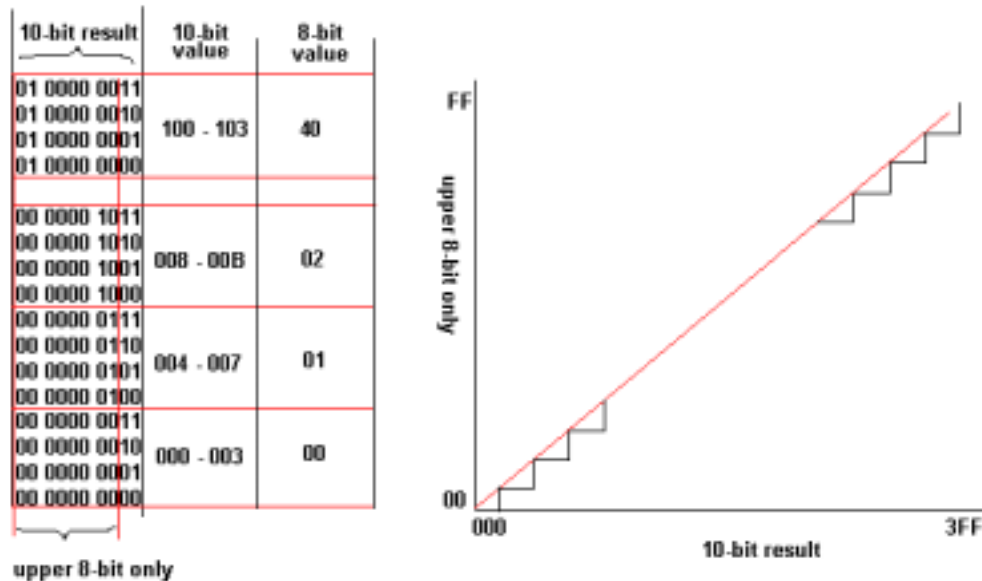


Fig. 59 Graph of FF against 3FF

However, except the resolution, it still can show some reasonable linear relationship of the result. As illustrated below, a value in the range 000 – 003h in 10-bit result would be just 00h in higher 8-bit only. Similarly, any value in the range of 008 – 00Bh in 10-bit result would be just 01h in the ADRESH only scheme with left justification. But coarse may be, the ADRESH only with the left justification still has the enough resolution power of analog value differentiation.

If you need very high resolution, you have to go with all 10-bit result. However, when high resolution is not needed, it is OK with the upper 8-bit result only. However, using all 10 bits is not difficult a matter in programming. It is only a matter of convenience or inconvenience. We will see the actual differences of the above two schemes with actual voltage reading.

The lowest 4 bits are allocated to decide the pin configuration, for analog pins or digital pins. For example, with PCFG3:PCFG0 = 0000, all the pins are assigned as analog input pins, i.e., A/D conversion pins. However, PCFG3:PCFG0=0110 or 0111 would make all the pins as digital I/O pins. Other combinations mix the analog and digital pins of the 8 channels. Another configuration included in these four bits is the selection of positive reference (V_{REF+}) and negative reference voltage (V_{REF-}) for A/D conversion. For example, PCFG3:PCFG0 = 0000 would select the logic power voltage (V_{DD} , or +5V, namely) as V_{REF+} and Ground, V_{SS} , as V_{REF-} . A reference voltage sets the maximum input voltage the A/D converter can convert. In other words, any voltage above the positive reference voltage, or any voltage below the negative reference voltage, would be saturated (or cut-off) to the reference voltage level. Therefore, with PCFG3:PCFG0 = 0000, any negative voltage would be treated as 0 volt or ground, and +5V is the maximum voltage can be converted. When we want to change the reference voltages, and

expand or shrink the voltage range of the analog input, we have to select appropriate combination of the 4 bits of ADCON1. For example, PCFG3:PCFG0 = 1000 allocates AN3 and AN2 for V_{REF+} and V_{REF-} , respectively.

The reference voltage, along with the number of bits used for conversion result, determines the step size of the converter, i.e., converter's resolution. For example, with positive reference voltage +5V and negative reference at the ground, the conversion range is 5V. This 5V is now divided by $2^{10} = 1024$ (maximum binary value of a 10-bit number) steps. Therefore, the step size is $5/1024 = 0.00488\text{V}$ or 4.88mV. Therefore the weight of each bit of the 10 bit results, therefore, has the multiple of the step voltage: bit 0 represents $2^0 \times 4.88\text{mV} = 4.88\text{mV}$; bit 1 represents $2^1 \times 4.88\text{mV} = 9.76\text{mV}$; bit 2 for $2^2 \times 4.88\text{mV} = 19.52\text{mV}$; and bit 9 for $2^9 \times 4.88\text{mV} = 2.5\text{V}$. Therefore, a result of 1000100010 would be interpreted as: $(2^9 + 2^5 + 2^1) \times 0.00488 = 2.664\text{V}$. As the equation shows, the maximum resolution we could get is 4.88mV. The resolution defines the smallest voltage change that can be measured. Every voltage below 4.88mV is read as 0 and any voltage above 4.88mV and below 9.76mV would be read as 4.88mV.

The last two registers involved in the A/D conversion are PIE1 (Peripheral Interrupt Enable 1) register and PIR1 (Peripheral Interrupt Request 1) register. PIE1 register is to grant or deny a peripheral interrupt and PIR1 register indicates the completion of a peripheral's process. From the both registers, we use only the 6th bit (ADIE from PIE1 and ADIF from PIR1) for A/D conversion control. Setting ADIF would trigger an interrupt whenever an A/D conversion is completed. Interrupt will be discussed later. In this chapter, we disable the interrupt for the time being. Clearing ADIE would not trigger an interrupt. Therefore, a completion of the A/D conversion should be checked by a "completion flag" bit. ADIF bit indicates the status of an A/D conversion process: ADIF=1 for completion and ADIF=0 for incompleteness. The completion flag bit must be cleared, after a completion of an A/D conversion, by software. Note that we do not use ADIF as an A/D conversion status bit, instead we use GO/~DONE bit of ADCON0 as the conversion status bit. ADIF bit is only to be cleared after GO/~DONE indicates the completion of A/D conversion.

PIE1 REGISTER (8Ch)

PSPIE	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE
-------	------	------	------	-------	--------	--------	--------

ADIE: A/D Converter Interrupt Enable bit
 1 = Enables the A/D converter interrupt
 0 = Disables the A/D converter interrupt

PIR1 REGISTER (0Ch)

PSPIF	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF
-------	------	------	------	-------	--------	--------	--------

ADIF: A/D Converter Interrupt Flag bit
 1 = An A/D conversion completed
 0 = The A/D conversion is not complete

Here goes the A/D conversion procedure in software perspective:

1. Make PORTA as inputs by setting all bits of TRISA register.
2. Disable A/D interrupt by clearing ADIE bit of PIE1 register.
3. Configure ADCON0 register.
4. Configure ADCON1 register.
5. Start A/D conversion by setting GO/ $\overline{\text{DONE}}$ bit of ADCON0 register.
6. Monitor GO/ $\overline{\text{DONE}}$ bit for a completion of the conversion. If the bit is cleared go to 7.
7. Conversion completed. Clear ADIF bit.
8. Move the content of ADRESH to a temporary space.
9. Move the content of ADRESL to another temporary space.

2. First Example of A/D Conversion

Let's have a simple voltage reading example with 16F877 by connecting a variable resistor between the +5V voltage source and the ground. Then connect the wiper terminal, which changes the terminal resistance and the terminal voltage from the ground, to AN0 channel of A/D conversion. We will read the voltage while changing the wiper position and display the value on a PC monitor. The first example is to display with two decimal point value for the voltage at the wiper terminal like. 2.50 or 1.96V by using only 8 MSBs of the result. The second example will use all 10 bit results and display with 3 decimal points like 2.496 or 1.962V.

Since we use only one channel (AN0) with the positive reference voltage and the negative reference voltage as +5 V and the ground, respectively, the configuration of ADCON0 goes like ADCON0=11000001 for internal RC clock, channel 0 (AN0), with A/D switch on. However, no conversion is started yet.

```
banksel    ADCON0                ;KKCCCGXO
movlw     0xC1                    ;11000001
movwf     ADCON0                  ;initialize ADC (RA0 is ADC port)
```

ADCON0 Register (1Fh)							
ADCS1	ADCS0	CHS2	CHS1	CHS0	GO/ $\overline{\text{DONE}}$	0	ADON
1	1	0	0	0	0	0	1
RC clock		Channel 0 (AN0)			A/D Stop		A/D turned on

For ADCON1, every bit is cleared to indicate "Left Justification" of the result and channel assignment along with reference voltage levels: ADCON1=00000000. We are going to ignore the 2 LSBs of the result stored in ADRESHL. Instead, we will take only ADRESH as if it comes from 8-bit A/D converter.

ADCON1 Register (9Fh)

		ADFM		PCFG3	PCFG2	PCFG1	PCFG0
1	1	0	0	0	0	0	1

Left Justified
All Analog Channel.
Vref+ = 5V
Vref- = 0V

```

movlw    0x00
banksel  ADCON1
movwf   ADCON1           ;PORT A is for ADC channel
                               ;With LEFT JUSTIFICATION
                               ;We will ignore two least significant
                               ;bits without much loss
    
```

The above lines are just a part of initialization. So now let's discuss about how to actually read and store the data, and then send to a PC for a display.

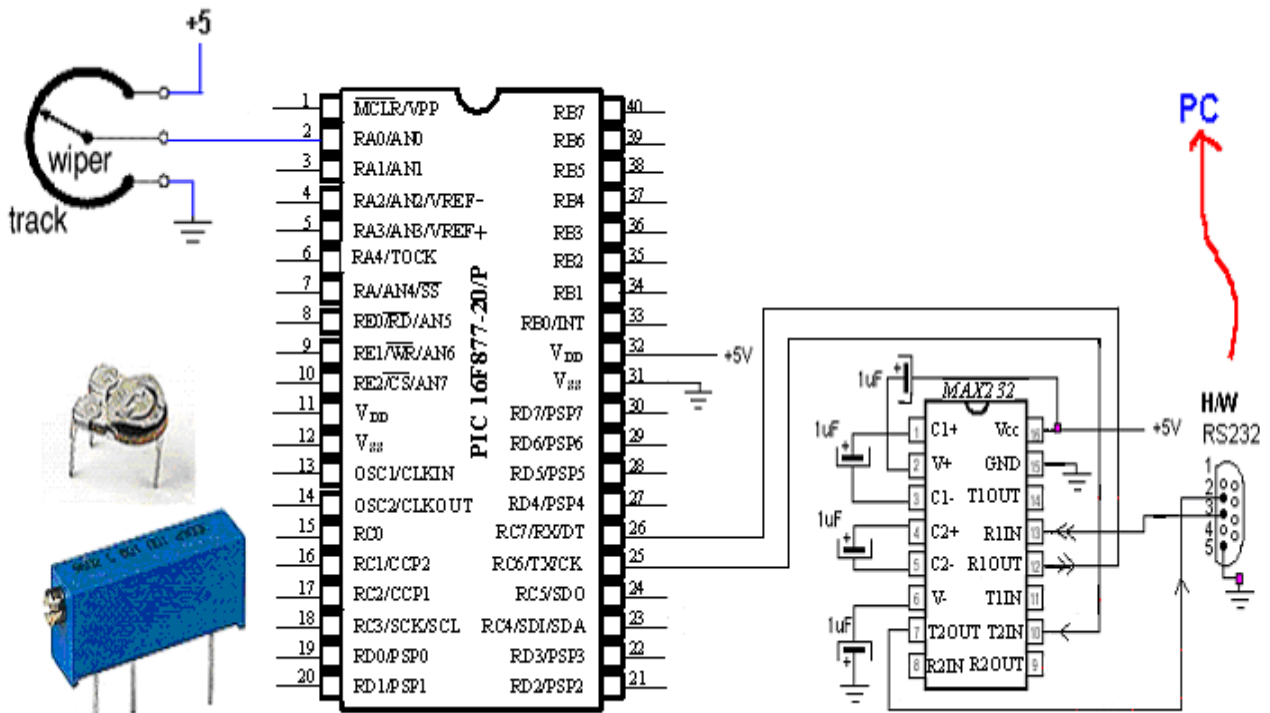


Fig. 60 A/D Conversion Example

Getting a conversion result would be much more convenient in a subroutine form since we are going to convert consecutively. The conversion routine starts from some delay to give the A/D module time to warm up. Then, we GO the conversion, and check the GO/~DONE bit is cleared indicating the completion of the conversion. When the conversion is finished, we clear ADIF bit, then, move the result in ADRESH to a temporary location.

```

;subroutine getADC
;The conversion result will be stored in W register
getADC
    
```

```

        call      delay10ms      ;warm up
        banksel  ADCON0
        bsf     ADCON0, GO      ;start conversion
ADCloop
        btfsc   ADCON0, GO      ;wait for completion
        goto    ADCloop
        bcf     PIR1, ADIF      ;clear conversion complete flag
        movf   ADRESH,0        ;store the result to W register
        return

```

Now, we have to determine the read value from AN0 by calling `getADC` subroutine. As we discussed in the section of A/D reference voltage, the 10-bit A/D conversion has, for [0, 5]V range, 4.88mV per conversion step. Therefore, the final measured voltage from the ADRESH can be simply formulated by:

$$V_{mea} = B_n \times 2^9 + B_n \times 2^8 + B_n \times 2^7 + B_n \times 2^6 + B_n \times 2^5 + B_n \times 2^4 + B_n \times 2^3 + B_n \times 2^2$$

where, B_n , $n = 0, \dots, 7$, are the bit values of ADRESH register ignoring the 2 LSBs in ADRESL.

The equation above looks too simple for a high-level language programmer, but it's not that simple in 16F877 programming. First, $2^9=512$ and $2^8=256$ are bigger than 1 byte value, which is the size of calculation and storage in 8-bit microcontroller. Of course, we can split the result into to registers, but still some burden we already feel. Second, after all the burdens we take for the bigger numbers, we still have problem to covert them into decimal point numbers. This problem is much bigger than the first one.

In 16F877 with Assembly language programming environment, it is much wiser to solve a problem by examining the bit pattern. Let's consider the 8-bit excepts as the 8-bit result. Then consider a value of the 8-bit result only when one bit is set.

```

1000 0000 =27=128
0100 0000 = 26= 64
0010 0000 =25=32
0001 0000 =24=16
0000 1000 =23=8
0000 0100 =22=4
0000 0010 =21=2
0000 0001 =20=1

```

Therefore, when all the bits are set, which occurs when the voltage reading is 5V, the numeric sum would be 255. Since our intention is to display in two decimal point format, we could say the value must correspond to 5.00. Let's name the single digit before the decimal point as D1 (digit 1), and two digits trailing the decimal point D2 (digit 2) and D3 (digit 3). But we can ignore the decimal point in the interpretation of the conversion result. Therefore, when we set the highest voltage as 500 instead of 5.00, there is no change or influence on the conversion result interpretation. The easiest way to convert the numerical values to 3 digit equivalent numbers so that the highest number corresponds to 500 is to double the numerical value for the 3-digit equivalent.


```

1000 0000 =27=128 → 256
0100 0000 = 26= 64 →128
0010 0000 =25=32 →064
0001 0000 =24=16 →032
0000 1000 =23=8 →016
0000 0100 =22=4 →008
0000 0010 =21=2 →004
0000 0001 =20=1 →002

```

However the sum of the 3 digit equivalents does not add up to 500, because the numerical sum is 255, not 250. The sum reaches at 510. So we need some minor massage around the numbers. What can be acceptable is shown as follows for a conversion of an 8-bit result to a two decimal point voltage value.

```

1000 0000 =27=128 → 250
0100 0000 = 26= 64 →125
0010 0000 =25=32 →063
0001 0000 =24=16 →032
0000 1000 =23=8 →016
0000 0100 =22=4 →008
0000 0010 =21=2 →004
0000 0001 =20=1 →002

```

Now, we check each bit of the 8-bit result stored in ADRESH. When the LSB is set, for example, the D3 must be increased by 2. If MSB is set, the D1 must be increased by 2. The highest digit D1 has its maximum at 5, so there is no reason to worry if the sum would be bigger 9. For example, an A/D conversion result is reached at ADRESH and its content is 00001110. Since we would have increased D3 by total 14 times, the value of D3 is 14 in decimal. This decimal number must be changed to 4 with carry 1 to the upper digit D2.

How do we automatically find if a sum is bigger than 9 and add the carry to the one upper digit? This procedure could be borrowed from BCD (Binary Coded Decimal) arithmetic. In BCD when sum of two BCD numbers (A digit BCD number occupies 4 binary bits) are bigger than 9, we add 6. For us, we would check the value of, say, D1, and if the value is equal to or greater than 10 (in decimal), we increase the one upper digit, D2, by one, subtract 10 from D1. If the value is 9 or less, we do not do anything at all. The same philosophy can be applied to D2. That means we can build a subroutine to check a value in a digit if it is below 10 or not. As we display 4 bit numbers from 7 – 15 as below, the numbers equal to or greater than 10 have: (bit3=1 and bit2=0) OR (bit3=1 and bit2=0 and bit1=1).

```

;subroutine to check >=10 or <10 =====
;W holds the value of a digit (D 1 or D2)
;HILO is a flag register to indicate the result
; >=10 ---> HILO=0
;<10 ---> HILO =0
TEN
    banksel    HILO

```

```

        clrf          HILO
        movwf        TEMPTEN          ;the content D1 (or D2) now to TEMPTEN
        btfss        TEMPTEN, 0x03    ;3rd bit check (8 or above)
        return       ;if third bit is zero, it is <10
        btfss        TEMPTEN, 0x02    ;if third bit is 1, check bit 2
        goto         nextbit
        bsf          HILO,0x00        ;if bit3=1 & bit2=1 it is >10
        return
nextbit
        btfss        TEMPTEN,0x01      ;if bit3=1 & bit2=0, then we check bit 1
        return
        bsf          HILO, 0x00        ;if bit3=1 &bit2=0 &bit1=1, it is >10
        return

```

The example code shown below displays the three digit voltage values on a monitor of a PC. In the example, a complete listing is provided. But I warn you here that the initialization of serial communication is not included in the code. In other words, the subroutine `Asynic_mode` which was discussed in Chapter 5 must be included, and it must be called at the very first part of the code. Otherwise, you do not get anything on your screen.

```

;ADC-V1.asm
;
;This program is to read voltage output from a rheostat
; and display the value on a PC terminal (current is updated every 2 seconds)
;
; AN0 is connected to the rheostat wiper
;;
;USE ONLY most significant 8 bits stored in ADRESH
;Max 5.00 V
;min 0.00 V
;
;PC's Hyper Terminal Set-Up: 8N1 19200
;Baud:      19200
;Data Bit:  8
;Parity:    None
;Stop Bit:  1
;Control:   None

```

```
list P = 16F877
```

```

STATUS    EQU    0x03
ZERO      EQU    0x02          ;Z flag
TXSTA     EQU    0x98          ;TX status and control
RCSTA     EQU    0x18          ;RX status and control
SPBRG     EQU    0x99          ;Baud Rate assignment
TXREG     EQU    0x19          ;USART TX Register
RCREG     EQU    0x1A          ;USART RX Register
PIR1      EQU    0x0C          ;USART RX/TX buffer status (empty or full)
RCIF      EQU    0x05          ;PIR1<5>: RX Buffer 1-Full 0-Empty
TXIF      EQU    0x04          ;PIR1<4>: TX Buffer 1-empty 0-full
TXMODE    EQU    0x20          ;TXSTA=00100000 : 8-bit, Async
RXMODE    EQU    0x90          ;RCSTA=10010000 : 8-bit, enable port, enable RX
BAUD      EQU    0x0F          ;0x0F (19200), 0x1F (9600)
PORTD     EQU    0x08
TRISD     EQU    0x88

```

```

PORTA      EQU    0x05
TRISA      EQU    0x85
ADCON0     EQU    0x1f
ADCON1     EQU    0x9f
ADRESH     EQU    0x1e      ;High Byte Result
ADRESL     EQU    0x9E      ;Low Byte Result
PIE1       EQU    0x8c
GO         EQU    0x02
ADIE       EQU    0x06
ADIF       EQU    0x06

;DISPLAY FORMAT (with two decimal points)
;
;-----  -----
;|      |      |      |      |
;|      |      |      |      |
;-----  -----
;

        CBLOCK      0x20
            temp
            tempten
            HIGHBYTE
            LOWBYTE
            HILO      ;flagging for 1(10 or bigger) or 0 (less than 10)
            DIGIT1
            DIGIT2
            DIGIT3      ;D1. D2 D3 (display format)
            ASCIIreg
            AD1
            AD2
            AD3      ;final 3 ASCII digits to be displayed
            Kount20us
            Kount120us ;Delay count (number of instr cycles for delay)
            Kount100us
            Kount1ms
            Kount10ms
            Kount1s
            Kount10s
            Kount1m
        ENDC

;=
;=====
        org          0x0000
        GOTO        START
        org          0x05
;=====

START
        movlw       0xFF
        banksel    TRISA
        movwf      TRISA      ;PORTA all inputs

        banksel    PIE1
        bcf        PIE1, ADIE ;disable ADC interrupt

```

```

    banksel    ADCON0            ;KKCCCGXO
    movlw     0xC1              ;11000001
    movwf    ADCON0            ;initialize ADC (RA0 is ADC port)

    movlw     0x00
    banksel    ADCON1
    movwf    ADCON1            ;PORT A is for ADC channel
                                ;With LEFT JUSTIFICATION
                                ;We will ignore two least significant
bits
                                ;without much loss
AGAIN
    banksel    PIR1
    bcf       PIR1, ADIF        ;clear conversion complete flag

    banksel    TEMP
    clrf     TEMP
    clrf     DIGIT1
    clrf     DIGIT2
    clrf     DIGIT3
    clrf     AD1                ;ASCII code for Digit1
    clrf     AD2                ;ASCII code for Digit2
    clrf     AD3                ;ASCII code for Digit3
    clrf     HILO

    call      GetADC            ;get the AD conversion result
    banksel    TEMP
    movwf    TEMP              ;Now TEMP holds the 8-bit ADC result
;
; No conversion to Max +5.00 Min 0.00 value
;pattern check
; 1000 0000 -->128 ---->250
; 0100 0000 -->64 ---->125
; 0010 0000 -->32 ----->063
; 0001 0000 -->16 ----->032
; 0000 1000 -->8 ----->016
; 0000 0100 -->4 ----->008
; 0000 0010 -->2 ----->004
; 0000 0001 -->1 ----->002

    movlw     0x00
B0    btfss   TEMP,0x00        ;check the bit 0 of the ADC result
    goto     B1
    incf     DIGIT3            ;DIGIT3=DIGIT3+2
    incf     DIGIT3

B1    btfss   TEMP, 0x01      ;bit 1 check
    goto     B2
    incf     DIGIT3
    incf     DIGIT3
    incf     DIGIT3
    incf     DIGIT3            ;Digit3=digit3+4

B2    btfss   TEMP, 0x02      ;bit 2 check
    goto     B3
    incf     DIGIT3

```

```

    incf    DIGIT3
    incf    DIGIT3
    incf    DIGIT3      ;
    incf    DIGIT3
    incf    DIGIT3
    incf    DIGIT3
    incf    DIGIT3      ;Digit3=digit3+8

; check if it is bigger than 10
    movf   DIGIT3, 0      ; to W
    call   TEN
    btfss  HILO,0x00
    goto   B3            ;Less than 10
    movlw  0x0A
    subwf  DIGIT3        ;f - 10 -->f
    incf   DIGIT2        ;Digit2=Digit2+1
    clrf   HILO

B3    btfss  TEMP, 0x03    ;bit 3 check
    goto   B4
    incf   DIGIT3
    incf   DIGIT3      ;
    incf   DIGIT3
    incf   DIGIT3
    incf   DIGIT3
    incf   DIGIT3      ;Digit3=digit3+6
    incf   DIGIT2      ;Digit2=Digit2+1

; check if it is bigger than 10
    movf   DIGIT3, 0      ; to W
    call   TEN
    btfss  HILO,0x00
    goto   B4            ;Less than 10
    movlw  0x0A
    subwf  DIGIT3
    incf   DIGIT2
    clrf   HILO

B4    btfss  TEMP, 0x04    ;bit 4 check
    goto   B5
    incf   DIGIT3
    incf   DIGIT3      ;Digit3=Digit3+2
    incf   DIGIT2
    incf   DIGIT2
    incf   DIGIT2      ;Digit2=Digit2+3

; check if it is bigger than 10
    movf   DIGIT3, 0      ; to W
    call   TEN
    btfss  HILO,0x00
    goto   B5            ;Less than 10
    movlw  0x0A
    subwf  DIGIT3
    incf   DIGIT2
    clrf   HILO

B5    btfss  TEMP, 0x05    ;bit 5 check
    goto   B6

```

```

    incf      DIGIT3
    incf      DIGIT3
    incf      DIGIT3      ;Digit3=Digit3+3
    incf      DIGIT2
    incf      DIGIT2
    incf      DIGIT2      ;
    incf      DIGIT2
    incf      DIGIT2
    incf      DIGIT2      ;Digit2=Digit2+6
; check if it is bigger than 10
    movf     DIGIT3, 0      ; to W
    call     TEN
    btfss    HILO,0x00
    goto     D2A      ;Less than 10
    movlw    0x0A
    subwf    DIGIT3
    incf     DIGIT2
    clrf     HILO
; Check DIGIT2 for 10 or above
D2A  movf     DIGIT2, 0
    call     TEN
    btfss    HILO, 0x00
    goto     B6
    movlw    0x0A
    subwf    DIGIT2
    incf     DIGIT1
    clrf     HILO

B6   btfss    TEMP, 0x06      ;bit 6 check
    goto     B7
    incf     DIGIT3
    incf     DIGIT3
    incf     DIGIT3
    incf     DIGIT3
    incf     DIGIT3      ;Digit3=Digit3+5
    incf     DIGIT2
    incf     DIGIT2      ;Digit2=Digit2+2
    incf     DIGIT1      ;Digit1=Digit1+1

; check if it is bigger than 10
    movf     DIGIT3, 0      ; to W
    call     TEN
    btfss    HILO,0x00
    goto     D2B      ;Less than 10
    movlw    0x0A
    subwf    DIGIT3
    incf     DIGIT2
    clrf     HILO
; Check DIGIT2 for 10 or above
D2B  movf     DIGIT2, 0
    call     TEN
    btfss    HILO, 0x00
    goto     B7
    movlw    0x0A
    subwf    DIGIT2
    incf     DIGIT1
    clrf     HILO

```

```

B7    btfss    TEMP, 0x07        ;bit 7 check
      goto    FINI
      incf    DIGIT2
      incf    DIGIT2
      incf    DIGIT2
      incf    DIGIT2
      incf    DIGIT2            ;Digit2=Digit2+5
      incf    DIGIT1
      incf    DIGIT1            ;Digit1=Digit1+2

; check if it is bigger than 10

      movf    DIGIT2, 0
      call   TEN
      btfss   HILO, 0x00
      goto   FINI
      movlw  0x0A
      subwf  DIGIT2
      incf   DIGIT1
      clrf   HILO

FINI
      movf    DIGIT1,0
      call   HTOA        ;ASCII conversion of Digit1
      movwf  AD1         ;final digit to be displayed
      movf    DIGIT2,0
      call   HTOA        ;ASCII conversion of Digit2
      movwf  AD2
      movf    DIGIT3,0
      call   HTOA        ;ASCII conversion of Digit3
      movwf  AD3

;
;ready to display
      movf    AD1, 0      ;First Digit
      call   TXPOLL
      movlw  '.'         ;Decimal Point
      call   TXPOLL
      movf    AD2,0
      call   TXPOLL      ;Second Digit
      movf    AD3,0
      call   TXPOLL      ;Third Digit
      call   CRLF        ;Line Change and Carriage Return
;delay 2s
      call   delay1s
      call   delay1s

      goto   AGAIN

;=====
;subroutine to check >=10 or <10 =====
; >=10 ---> HILO=0
; <10 --->HILO =0

```

```

TEN
    banksel    HILO
    clrf       HILO
    movwf      TEMPTEN
    btfss      TEMPTEN, 0x03    ;3rd bit
    return
    btfss      TEMPTEN, 0x02
    goto       nextbit
    bsf        HILO,0x00
    return
nextbit
    btfss      TEMPTEN,0x01
    return
    bsf        HILO, 0x00
    return
;-----
;Subroutine GetADC =====
getADC
    call       delay10ms    ;warm up
    banksel    ADCON0
    bsf        ADCON0, GO    ;start conversion
ADCloop
    btfsc      ADCON0, GO    ;wait for conversion to finish
    goto       ADCloop
    bcf        PIR1, ADIF    ;clear conversion complete flag
    movf       ADRESH,0
    return
;-----
;RS232 TX subroutine =====
TXPOLL
    banksel    PIR1
    btfss      PIR1, TXIF    ; Check if TX buffer is empty
    goto       TXPOLL
    banksel    TXREG
    movwf      TXREG          ; Place the character to TX buffer
    return
;-----
;To send CR and LF =====
CRLF
    movlw     H'0d'          ;CR
    call      TXPOLL
    movlw     H'0a'          ;LF
    call      TXPOLL
    return
;-----
;; === hex to ascii conversion subroutine
;move the content to W before call this routine
;final result will be stored back to W
HTOA
    movwf     ASCIIreg
;check 0-9 or A-F
    btfsc     ASCIIreg, 0x03    ;0 - 7
    goto     RECHK
THIRTY
    movlw     0x30
    addwf     ASCIIreg
    movf      ASCIIreg,0

```



```

        return

RECHK andlw      0x06 ;
        btfsc    STATUS, ZERO
        goto     THIRTY
        movlw    0x37
        addwf    ASCIIreg
        movf     ASCIIreg, 0
        return

;-----
;DELAY SUBROUTINES

Delay20us
        banksel  Kount20us
        movlw    H'1F' ;D'31'
        movwf    Kount20us
R20us  decfsz   Kount20us
        goto     R20us
        return

;
;
Delay120us
        banksel  Kount120us
        movlw    H'C5' ;D'197'
        movwf    Kount120us
R120us decfsz   Kount120us
        goto     R120us
        return

;
;
Delay100us
        banksel  Kount100us
        movlw    H'A4'
        movwf    Kount100us
R100us decfsz   Kount100us
        goto     R100us
        return

;
;10ms delay
; call 100 times of 100 us delay (with some time discrepancy)
Delay10ms
        banksel  Kount10ms
        movlw    H'64' ;100
        movwf    Kount10ms
R10ms  call     delay100us
        decfsz   Kount10ms
        goto     R10ms
        return

;
;1 sec delay
;call 100 times of 10ms delay
Delay1s
        banksel  Kount1s
        movlw    H'64'
        movwf    Kount1s

```

```

R1s    call        Delay10ms
        decfsz     Kount1s
        goto       R1s
        return

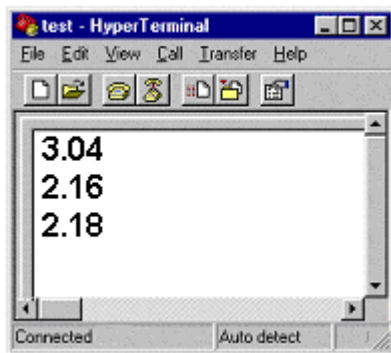
;
;
;10 s delay
;call 10 times of 1 s delay
Delay10s
        banksel   Kount10s
        movlw     H'0A'    ;10
        movwf    Kount10s
R10s   call        Delay1s
        decfsz     Kount10s
        goto       R10s
        return

;
;1 min delay
;call 60 times of 1 sec delay
Delay1m
        banksel   Kount1m
        movlw     H'3C'    ;60
        movwf    Kount1m
R1m    call        Delay1s
        decfsz     Kount1m
        goto       R1m
        return

;=====
                END
;END OF PROGRAM

```

How do you feel about this rather a long line of code for just displaying a simple number? As we see most of the code are devoted to interpretation and display, rather than A/D conversion itself. If you do not have to display the measured voltage, but instead compare with a threshold value, code would be much shorter and simpler. Anyway, see if you have the result like illustrated below as you change the wiper position of the variable resistor.



Now let's use all 10-bit result for the same variable resistor setting we used for 8-bit result calculation. With a similar pattern observation and interpretation, we can relate the 10-bit result to three decimal digit voltage. The maximum voltage (without displaying the decimal point) would be, then, 5000.

10 0000 0000	= 2 ⁹ = 512	→2500
01 0000 0000	=2 ⁸ =256	→1250
00 1000 0000	=2 ⁷ =128	→ 0625
00 0100 0000	=2 ⁶ = 64	→0312
00 0010 0000	=2 ⁵ =32	→0158
00 0001 0000	=2 ⁴ =16	→0080
00 0000 1000	=2 ³ =8	→0040
00 0000 0100	=2 ² =4	→0020
00 0000 0010	=2 ¹ =2	→0010
00 0000 0001	=2 ⁰ =1	→0005

The only difference in the 3 decimal digit case is that we have to have one more digit value D4 (or digit 4) and its ASCII equivalent (AD4). The subroutines for A/D conversion, check for a digit value if it is below 10 or not, hex to ASCII conversion, and time delay are all the same, except a minor change in the `getADC` subroutine, since we have to store the values of `ADRESH` and `ADRESL`. So slightly revised subroutine, `getADC2`, is shown below.

```

;Subroutine GetADC2 =====
getADC2
    call    delay10ms    ;warm up
    banksel ADCON0
    bsf    ADCON0, GO    ;start conversion
ADCloop btfscl ADCON0, GO    ;wait for conversion to finish
    goto   ADCloop
    bcf    PIR1, ADIF    ;clear conversion complete flag
    movf   ADRESH, 0
    banksel tempHIGH
    movwf  tempHIGH      ;tempHIGH <--ADRESH
    banksel ADRESL
    movf   ADRESL, 0
    banksel tempLOW
    movwf  tempLOW      ;tempLOW <--ADRESL
    return

```

Another slight change in the code is the justification of the A/D conversion result: we select this time "Right Justification" so that lower 8 bits are stored in `ADRESL` and the upper 2 bits of the results are stored at the lowest 2 LSB positions of `ADRESH`. The example code, without subroutines, is displayed below.

```

;ADC-V2.asm
;
;This program is to read voltage output from a rheostat
; and display the value on a PC terminal (current is updated every 2 seconds)
;
; AN0 is connected to the rheostat wiper
;
;USE whole 10 bits
;MAX 5.000 V

```

```

;Min 0.000 V
;PC's Hyper Terminal Set-Up: 8N1 19200
;Baud:      19200
;Data Bit:  8
;Parity:    None
;Stop Bit:  1
;Control:   None

        list P = 16F877

STATUS      EQU    0x03
ZERO        EQU    0x02        ;Z flag
TXSTA       EQU    0x98        ;TX status and control
RCSTA       EQU    0x18        ;RX status and control
SPBRG       EQU    0x99        ;Baud Rate assignment
TXREG       EQU    0x19        ;USART TX Register
RCREG       EQU    0x1A        ;USART RX Register
PIR1        EQU    0x0C        ;USART RX/TX buffer status (empty or full)
RCIF        EQU    0x05        ;PIR1<5>: RX Buffer 1-Full 0-Empty
TXIF        EQU    0x04        ;PIR1<4>: TX Buffer 1-empty 0-full
TXMODE      EQU    0x20        ;TXSTA=00100000 : 8-bit, Async
RXMODE      EQU    0x90        ;RCSTA=10010000 : 8-bit, enable port, enable RX
BAUD        EQU    0x0F        ;0x0F (19200), 0x1F (9600)
PORTD       EQU    0x08
TRISD       EQU    0x88
PORTA       EQU    0x05
TRISA       EQU    0x85
ADCON0      EQU    0x1f
ADCON1      EQU    0x9f
ADRESH      EQU    0x1e        ;High Byte Result
ADRESL      EQU    0x9E        ;Low Byte Result
PIE1        EQU    0x8c
GO          EQU    0x02
ADIE        EQU    0x06
ADIF        EQU    0x06

;DISPLAY FORMAT (with three decimal points)
;
; ----  ----  ----  ----
; |  |  |  |  |  |  |  |
; |  |  |  |  |  |  |  |
; ----  ----  ----  ----
;
;
        CBLOCK      0x20
                tempHIGH      ;storage space of ADC result
                tempLOW
                tempTEN
                HILO          ;flagging for 1(10 or bigger) or 0 (less than 10)
                DIGIT1
                DIGIT2
                DIGIT3
                DIGIT4      ;D1. D2 D3 (display format) Double precision
                ASCIIreg
                AD1

```

```

        AD2
        AD3
        AD4          ;final 4 ASCII digits to be displayed
        Kount20us
        Kount120us   ;Delay count (number of instr cycles for delay)
        Kount100us
        Kount1ms
        Kount10ms
        Kount1s
        Kount10s
        Kount1m
    ENDC
;
;=====
        org          0x0000
        GOTO         START
        org          0x05
;=====

START
    movlw           0xFF
    banksel        TRISA
    movwf          TRISA          ;PORTA all inputs

    banksel        PIE1
    bcf            PIE1, ADIE     ;disable ADC interrupt

    banksel        ADCON0
    movlw          0xC1
    movwf          ADCON0        ;initialize ADC (RA0 is ADC port)

    movlw          0x80          ;
    banksel        ADCON1
    movwf          ADCON1        ;PORT A is for ADC channel
                                ;With RIGHT JUSTIFICATION
                                ;ADRESH(B9 and B8)
                                ;ADRESL (B7 - B0)

AGAIN
    banksel        PIR1
    bcf            PIR1, ADIF     ;clear conversion complete flag

    banksel        TEMPHIGH
    clrf          TEMPHIGH
    clrf          TEMPLOW
    clrf          DIGIT1
    clrf          DIGIT2
    clrf          DIGIT3
    clrf          DIGIT4
    clrf          AD1
    clrf          AD2
    clrf          AD3
    clrf          AD4
    clrf          HILO

    call          GetADC2

```

```

;Now tempLOW holds the lower 8-bit ADC result
;tempHIGH for the upper 2 bits

;
; Conversion to Max +5.000 Min 0.000 value
; 98 7654 3210 (bit)
; 10 0000 0000 -->512 ---->2500
; 01 0000 0000 -->256 ---->1250
; 00 1000 0000 -->128 ---->0625
; 00 0100 0000 -->64 ---->0312
; 00 0010 0000 -->32 ----->0158
; 00 0001 0000 -->16 ----->0080
; 00 0000 1000 -->8 ----->0040
; 00 0000 0100 -->4 ----->0020
; 00 0000 0010 -->2 ----->0010
; 00 0000 0001 -->1 ----->0005

        movlw      0x00
        banksel   tempLOW
B0      btfss     TEMPLOW,0x00      ;check the bit 0 of the ADC result
        goto      B1
        incf     DIGIT4             ;DIGIT4=DIGIT4+5
        incf     DIGIT4
        incf     DIGIT4
        incf     DIGIT4
        incf     DIGIT4

B1      btfss     TEMPLOW, 0x01     ;bit 1 check
        goto      B2
        incf     DIGIT3             ;Digit3=digit3+1

B2      btfss     TEMPLOW, 0x02     ;bit 2 check
        goto      B3
        incf     DIGIT3
        incf     DIGIT3             ;Digit3=digit3+2

B3      btfss     TEMPLOW, 0x03     ;bit 3 check
        goto      B4
        incf     DIGIT3
        incf     DIGIT3
        incf     DIGIT3
        incf     DIGIT3             ;Digit3=digit3+4

B4      btfss     TEMPLOW, 0x04     ;bit 4 check
        goto      B5
        incf     DIGIT3
        incf     DIGIT3
        incf     DIGIT3
        incf     DIGIT3
        incf     DIGIT3
        incf     DIGIT3
        incf     DIGIT3             ;DIGIT3=DIGIT3+8

; check if it is bigger than 10
        movf     DIGIT3, 0      ; to W
        call     TEN
        btfss     HILO,0x00
        goto     B5             ;Less than 10

```

```

        movlw      0x0A
        subwf     DIGIT3
        incf      DIGIT2
        clrf      HILO

B5      btfss     TEMLOW, 0x05      ;bit 5 check
        goto     B6
        incf     DIGIT4
        incf     DIGIT4
        incf     DIGIT4
        incf     DIGIT4
        incf     DIGIT4
        incf     DIGIT4
        incf     DIGIT4
        incf     DIGIT4      ;Digit4=Digit4+8
        incf     DIGIT3
        incf     DIGIT3
        incf     DIGIT3
        incf     DIGIT3
        incf     DIGIT3      ;Digit3=Digit3+5
        incf     DIGIT2      ;Digit2=Digit2+1
; check if it is bigger than 10
        movf     DIGIT4, 0      ; to W
        call    TEN
        btfss   HILO, 0x00
        goto    D5A      ;Less than 10
        movlw   0x0A
        subwf   DIGIT4
        incf    DIGIT3
        clrf    HILO
; Check DIGIT2 for 10 or above
D5A     movf    DIGIT3, 0
        call    TEN
        btfss   HILO, 0x00
        goto    B6
        movlw   0x0A
        subwf   DIGIT3
        incf    DIGIT2
        clrf    HILO

B6      btfss     TEMLOW, 0x06      ;bit 6 check
        goto     B7
        incf     DIGIT4
        incf     DIGIT4      ;digit4=digit4+2
        incf     DIGIT3      ;Digit3=Digit3+1
        incf     DIGIT2
        incf     DIGIT2
        incf     DIGIT2      ;Digit2=Digit2+3

; check if it is bigger than 10
        movf     DIGIT4, 0      ; to W
        call    TEN
        btfss   HILO, 0x00
        goto    D6A      ;Less than 10
        movlw   0x0A
        subwf   DIGIT4

```

```

        incf      DIGIT3
        clrfs    HILO
; Check DIGIT2 for 10 or above
D6A    movf      DIGIT3, 0
        call     TEN
        btfss   HILO, 0x00
        goto    B7
        movlw   0x0A
        subwf   DIGIT3
        incf    DIGIT2
        clrfs    HILO

B7     btfss   TEMLOW, 0x07    ;bit 7 check
        goto    B8
        incf    DIGIT4
        incf    DIGIT4
        incf    DIGIT4
        incf    DIGIT4
        incf    DIGIT4    ;digit4=digit4+5
        incf    DIGIT3
        incf    DIGIT3    ;digit3=digit3+2
        incf    DIGIT2
        incf    DIGIT2
        incf    DIGIT2
        incf    DIGIT2
        incf    DIGIT2
        incf    DIGIT2
        incf    DIGIT2    ;Digit2=Digit2+6

; check if it is bigger than 10

        movf    DIGIT4, 0
        call    TEN
        btfss   HILO, 0x00
        goto    D7A
        movlw   0x0A
        subwf   DIGIT4
        incf    DIGIT3
        clrfs    HILO

; check if it is bigger than 10

D7A    movf    DIGIT3, 0
        call    TEN
        btfss   HILO, 0x00
        goto    D7B
        movlw   0x0A
        subwf   DIGIT3
        incf    DIGIT2
        clrfs    HILO
; check if it is bigger than 10

D7B    movf    DIGIT2, 0
        call    TEN
        btfss   HILO, 0x00
        goto    B8

```



```

        movlw      0x0A
        subwf     DIGIT2
        incf      DIGIT1
        clrf      HILO

B8      btfss     TEMPHIGH, 0x00    ;bit 8 check
        goto     B9
        incf     DIGIT3
        incf     DIGIT3
        incf     DIGIT3
        incf     DIGIT3
        incf     DIGIT3          ;digit3=digi3+5
        incf     DIGIT2
        incf     DIGIT2          ;Digit2=Digit2+2
        incf     DIGIT1          ;digit1=digit1+1

; check if it is bigger than 10

        movf     DIGIT3, 0
        call    TEN
        btfss   HILO, 0x00
        goto    D8A
        movlw   0x0A
        subwf  DIGIT3
        incf   DIGIT2
        clrf   HILO

; check if it is bigger than 10

D8A     movf     DIGIT2, 0
        call    TEN
        btfss   HILO, 0x00
        goto    B9
        movlw   0x0A
        subwf  DIGIT2
        incf   DIGIT1
        clrf   HILO

B9      btfss     TEMPHIGH, 0x01    ;bit 9 check
        goto     FINI
        incf     DIGIT2
        incf     DIGIT2
        incf     DIGIT2
        incf     DIGIT2
        incf     DIGIT2          ;Digit2=Digit2+5
        incf     DIGIT1
        incf     DIGIT1          ;digit1=digit1+2

; check if it is bigger than 10

        movf     DIGIT2, 0
        call    TEN
        btfss   HILO, 0x00
        goto    FINI

```

```

        movlw      0x0A
        subwf     DIGIT2
        incf     DIGIT1
        clrf     HILO

FINI
        movf     DIGIT1,0
        call    HTOA
        movwf   AD1           ;final digit to be displayed
        movf     DIGIT2,0
        call    HTOA
        movwf   AD2
        movf     DIGIT3,0
        call    HTOA
        movwf   AD3
        movf     DIGIT4,0
        call    HTOA
        movwf   AD4

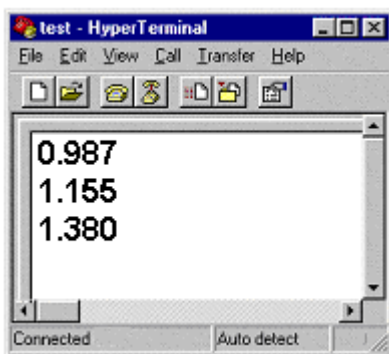
;
;ready to display
        movf     AD1, 0
        call    TXPOLL
        movlw   '.'
        call    TXPOLL
        movf     AD2,0
        call    TXPOLL
        movf     AD3,0
        call    TXPOLL
        movf     AD4,0
        call    TXPOLL
        call    CRLF  ;Line Change

;delay 2s
        call    delay1s
        call    delay1s

        goto   AGAIN

```

As you run the code while changing the wiper position of the variable resistor, we expect to see the following or similar display.



3. A/D Application to Infrared Ranger for Distance Measurement

IR ranger is a general purpose distance measuring sensor which usually consists of IR emitting diodes, position sensitive detector, and signal processing circuit. Basically it measures the distance by the time elapsed between an IR transmission and IR reception.

SHARP's **GP2D12** sensor takes distance reading and reports the distance as an analog voltage with a distance range of 10cm (~4") to 80cm (~30"). The interface is 3-wire with power, ground and the output voltage and requires a JST 3-pin connector which is included with each detector package. This is a common, robust, inexpensive sensor.

As the Distance vs. Voltage curve shows the output voltage is gradually decreased as the distance increases. So even though the specification says that the maximum distance the GP2D12 can measure is 80cm, we could extend the range further, since the voltage further reduces as the distance increases. The big problem of this ranger is that there is one discontinuity point: below 10cm the voltage change is reversed to decrease. Therefore, when you have, say, 2.8 V, you are not sure whether the distance is 15cm or 5 cm. When you use this ranger as many do, you have to be very careful that your application platform, robot or vehicle, should not approach an obstacle too close, less than 10 cm. Easiest solution is to give much more clearance from the obstacle, like 30cm, and if the output voltage from the range further increases, then you back off your robot or vehicle.

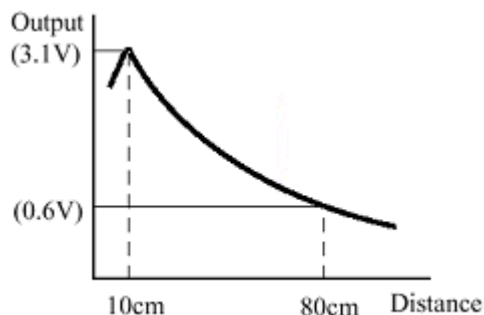


Fig. 61(a) Distance vs. Voltage curve



Fig. 61(b) SHARP GP2D12 Detector

As described, the ranger application is just another example of A/D conversion. Since the maximum voltage is less than +5V, we use the same configuration we used for the variable resistor. Except that we are going to connect the GP2D12 to AN1, instead. Since the voltage-distance relationship is nonlinear, we have to have a kind of table to interpret the voltage we get from A/D conversion to actual distance between the ranger and an obstacle. We will apply the same 8-bit result only approach for this example. Also we will display the distance on a PC monitor. The distance display format is with 3 digits.

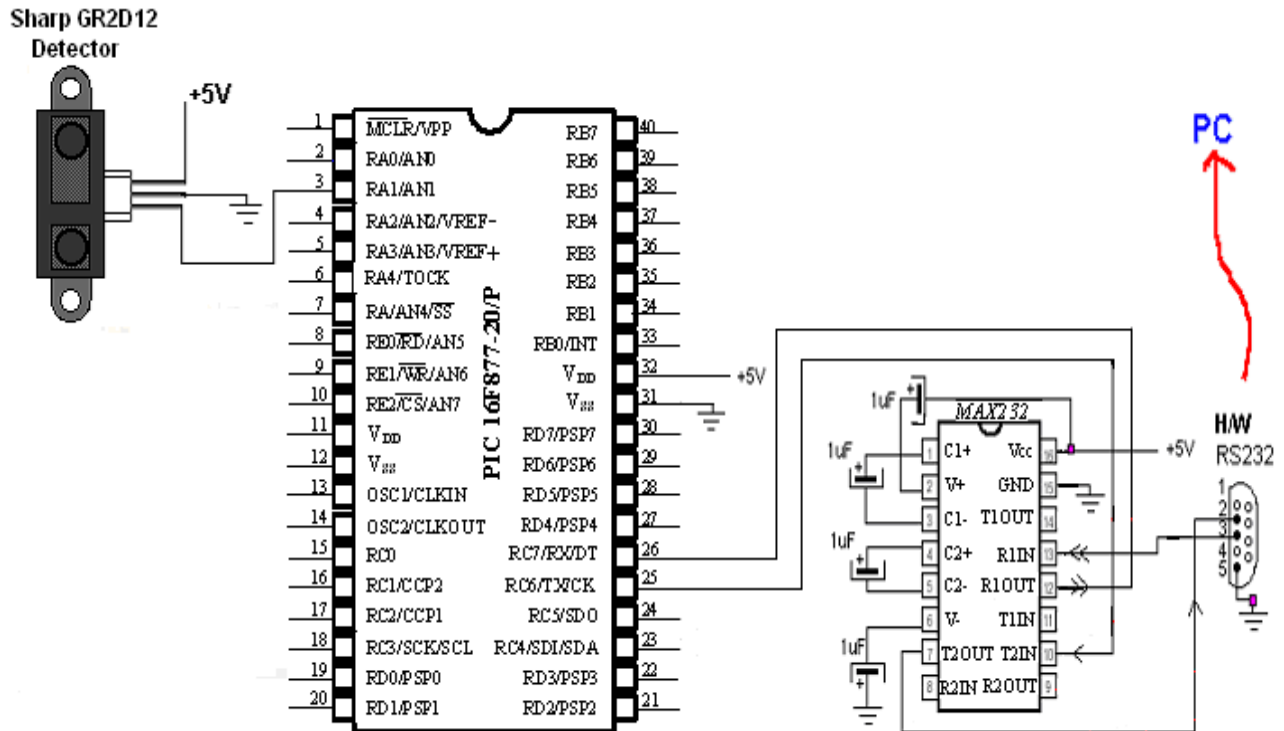


Fig. 62 Sharp GR2D12 Detector connection to PIC 16F877

For the conversion, we use the same logic we used for the variable resistance and the wiper voltage measurement. However the sum of the 3 digit equivalents does not add up to 500, because the numerical sum is 255, not 250. The sum reaches at 510. So we need some minor massage around the numbers. What can be acceptable is shown as follows for a conversion of an 8-bit result to a two decimal point voltage value.

- 1000 0000 = $2^7=128$ → 2.50 [V]
- 0100 0000 = $2^6= 64$ →1.25 [V]
- 0010 0000 = $2^5=32$ →0.63 [V]
- 0001 0000 = $2^4=16$ →0.32 [V]
- 0000 1000 = $2^3=8$ →0.16 [V]
- 0000 0100 = $2^2=4$ →0.08 [V]
- 0000 0010 = $2^1=2$ →0.04 [V]
- 0000 0001 = $2^0=1$ →0.02 [V]

From the test of the output voltage vs. distance using an oscilloscope and an obstacle, we found that nonlinear relationship of voltage and current as follows.

Distance [cm]	10	20	40	50	60	80	100	>100
Output Voltage [V]	2.4	1.25	0.7	0.58	0.5	0.42	0.34	<0.33

For 2.4 V for 10cm, since we are not aiming for very accurate (actually we cannot do that with

the ranger), we use the value of 1.25V as an approximate value with the bit pattern of 10000000. Also, since the 10cm is the closest distance measurement, higher than this value can be ignore. For 1.25V for 20cm matches well with the bit pattern of 01000000. The 0.7 V for 40cm can also be approximated by 0.63V or bit pattern of 00100000. However, making 0.58V for 50cm is impossible by just one bit information. Instead, since $0.58 = 0.32 + 0.16 + 0.08$, we can use the three bit information for 40 cm: 00011100 as a conclusion. In a similar manner, we can have the following simple distance interpretation pattern from the output voltage of the ranger.

```

1xxx xxxx -->128 --->250----->010cm
01xx xxxx -->64 --->125----->020cm
001x xxxx -->32 ---->063----->040cm
0001 11xx -->28 ---->056----->050cm
0001 10xx -->24 ---->048----->060cm
0001 01xx -->20 ---->040----->080cm
0001 00xx -->16 ---->032----->100cm
0000 xxxx -->8 ---->016 ----->Out of Range

```

The Pseudo-Code for voltage to distance interpretation is shown below. The label 'cm' followed by a three digit number indicates the place for displaying the distance of the number. The label 'cmqqq' is for displaying out of range distance situation.

```

                if B7=1 goto cm010, else goto B           ;<B7>=1 for 10cm
B:              if B6=1 goto cm020, else goto C           ;<B7:B6>=01 for 20cm
C:              if B5=1 goto cm040, else goto D           ;<B7:B5>=001 for 40cm
D:              if B4=1 goto D1, else goto E
                D1:   if B3=1 goto D2, else goto D3
                D2:   if B2=1 goto cm050, else goto cm060 ;<B7:B2>=000111 for 50cm
                                                         ;<B7:B2>=000110 for 60cm
                D3:   if B2=1 goto cm080, else goto cm100 ;<B7:B2>=000101 for 80cm
                                                         ;<B7:B2>=000100 for 100cm
E:              goto cmqqq                               ;<B7:B4>=0000 for Out of Range

```

The following example code is a full program for Sharp GP2D12 without listing subroutine. The subroutines needed this code are the same ones we used in the first example of A/D conversion reading the wiper voltage from the variable resistor.

```

;GP2D12.asm
;
;This program is to read voltage output from
; a Sharp Ranger GP2D12D
; and display the value on a PC terminal (updated every half second)
;
; AN1 is connected to the Ranger
;;
;USE ONLY most significant 8 bits
;Max 5.00 V
;min 0.00 V
;

```

```

list P = 16F877

STATUS      EQU    0x03
ZERO        EQU    0x02      ;Z flag
TXSTA       EQU    0x98      ;TX status and control
RCSTA       EQU    0x18      ;RX status and control
SPBRG       EQU    0x99      ;Baud Rate assignment
TXREG       EQU    0x19      ;USART TX Register
RCREG       EQU    0x1A      ;USART RX Register
PIR1        EQU    0x0C      ;USART RX/TX buffer status (empty or full)
RCIF        EQU    0x05      ;PIR1<5>: RX Buffer 1-Full 0-Empty
TXIF        EQU    0x04      ;PIR1<4>: TX Buffer 1-empty 0-full
TXMODE      EQU    0x20      ;TXSTA=00100000 : 8-bit, Async
RXMODE      EQU    0x90      ;RCSTA=10010000 : 8-bit, enable port, enable RX
BAUD        EQU    0x0F      ;0x0F (19200), 0x1F (9600)
PORTD       EQU    0x08
TRISD       EQU    0x88
PORTA       EQU    0x05
TRISA       EQU    0x85
ADCON0      EQU    0x1f
ADCON1      EQU    0x9f
ADRESH      EQU    0x1e      ;High Byte Result
ADRESL      EQU    0x9E      ;Low Byte Result
PIE1        EQU    0x8c
GO          EQU    0x02
ADIE        EQU    0x06
ADIF        EQU    0x06

;DISPLAY FORMAT
;
;XXX CM (3 digit)
;
;
        CBLOCK      0x20
            temp
            tempten
            HIGHBYTE
            LOWBYTE
            HILO      ;flagging for 1(10 or bigger) or 0 (less than 10)
            DIGIT1
            DIGIT2
            DIGIT3    ;D1. D2 D3 (display format) Double precision
            ASCIIreg
            AD1
            AD2
            AD3      ;final 3 ASCII digits to be displayed
            Kount20us
            Kount120us ;Delay count (number of instr cycles for delay)
            Kount100us
            Kount1ms
            Kount10ms
            Kount100ms
            Kount500ms
            Kount1s
            Kount10s

```

```

        Kount1m
        ENDC
;
;The Next
;Bootloader first execute the first 4 addresses
;=====
        org      0x0000
        goto    START
        org      0x05
;=====
START
        movlw   0xFF
        banksel TRISA
        movwf   TRISA      ;PORTA all Inputs

        banksel PIE1
        bcf     PIE1, ADIE ;disable ADC interrupt

        banksel ADCON0      ;KKCCCGXO
        movlw   0xC9        ;11001001
        movwf   ADCON0      ;initialize ADC (AN1 is ADC port)

        movlw   0x00
        banksel ADCON1
        movwf   ADCON1      ;PORT A is for ADC channel
                                ;With LEFT JUSTIFICATION
                                ;We will ignore two least significant
bits
                                ;without much loss
AGAIN
        banksel PIR1
        bcf     PIR1, ADIF  ;clear conversion complete flag

        banksel TEMP
        clrf   TEMP
        clrf   AD1
        clrf   AD2
        clrf   AD3

        call   GetADC      ;voltage reading
        banksel TEMP
        movwf  TEMP        ;Now TEMP holds the 8-bit ADC result
        movlw  '0'
        movwf  AD3         ;AD3=0
        movwf  AD1         ;AD2=0
        movwf  AD2         ;Ad1=0
                                ;Distance = 000 now
;
; Max +5.00 Min 0.00
;Conversion to Distance (See the nonlinear graph for GP2D12)
;Experimental results
;
; 2.70 [V]   9 cm
; 2.4         10
; 1.25         20
; 1.1         25

```

```

;0.9      30
;0.8      35
;0.7      40
;0.62     45
;0.58     50
;0.52     55
;0.5      60
;0.48     65
;0.46     70
;0.44     75
;0.42     80
;0.4      85
;0.38     90
;0.36     95
;0.34     100
;0.32     110
;0.3      140
;0.28     150

;FROM ADC with LEFT JUSTIFIED
; 1 0000 0000  -->256  ---->500
; 0 1xxx xxxx  -->128  ---->250----->010cm
; 0 01xx xxxx  -->64   ---->125----->020cm (B)
; 0 001x xxxx  -->32   ---->063----->040cm (C)
; 0 0001 11xx  -->28   ---->056----->050cm (D, D1)
; 0 0001 10xx  -->24   ---->048----->060cm (D2)
; 0 0001 01xx  -->20   ---->040----->080cm (D3)
; 0 0001 00xx  -->16   ---->032----->100cm (D3)
; 0 0000 xxxx  -->8    ---->016 ---->Out of Range (E)

;Pseudo-Code
;
; if B7=1 goto cm010, else goto B
; B: if B6=1 goto cm020, else goto C
; C: if B5=1 goto cm040, else goto D
; D: if B4=1 goto D1, else goto E
;   D1: if B3=1 goto D2, else goto D3
;   D2: if B2=1 goto cm050, else goto cm060
;   D3: if B2=1 goto cm080, else goto cm100
; E: goto cmqqq

          btfss    TEMP, 0x07
          goto     BB
          goto     cm010
BB        btfss    TEMP, 0x06
          goto     CC
          goto     cm020
CC        btfss    TEMP, 0x05
          goto     DD
          goto     cm040

DD        btfss    TEMP, 0x04
          goto     EE
D1        btfss    TEMP, 0x03
          goto     D3

```



```

        goto        D2

D2      btfss      TEMP, 0x02
        goto        cm060
        goto        cm050

D3      btfss      TEMP, 0x02
        goto        cm100
        goto        cm080

EE      goto        cmqqq          ;end of interpretation

cm010  movlw      '1'
        movwf      AD2            ;Distance = 010
        goto        FINI
cm020  movlw      '2'
        movwf      AD2            ;Distance = 020
        goto        FINI
cm040  movlw      '4'
        movwf      AD2            ;040
        goto        FINI
cm050  movlw      '5'
        mmovwf     AD2            ;050
        goto        FINI
cm060  movlw      '6'
        movwf      AD2            ;060
        goto        FINI
cm080  movlw      '8'
        movwf      AD2            ;080
        goto        FINI
cm100  movlw      '1'
        movwf      AD1            ;100
        goto        FINI
cmqqq  movlw      '>'
        movwf      AD1
        movwf      AD2
        movwf      AD3            ;>>> to indicate out-of-range

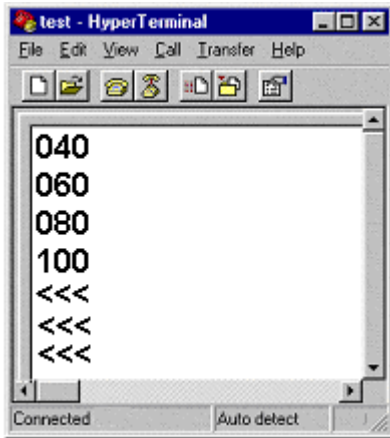
FINI

;ready to display
        movf       AD1, 0
        call       TXPOLL
        movf       AD2, 0
        call       TXPOLL
        movf       AD3, 0
        call       TXPOLL
        call       CRLF          ;Line Change
;delay 2s
        call       delay100ms

        goto        AGAIN        ;repeat every 100ms

```

We expect to see the following display when we move an object away from the ranger.



4. Current Measurement Applications using A/D converter Module

As the last example of A/D conversion in data acquisition, a few current sensors are introduced here. Some are good for smaller current and others are better for rather a larger current. A simple thermister is also introduced for a simple temperature measurement using the A/D module.

CSA-1V hall effect current sensor measures a current through itself by the principle of the magnetic field generated by the current flow. The amount of the current will be translated by the chip and the corresponding value in voltage will be produced.

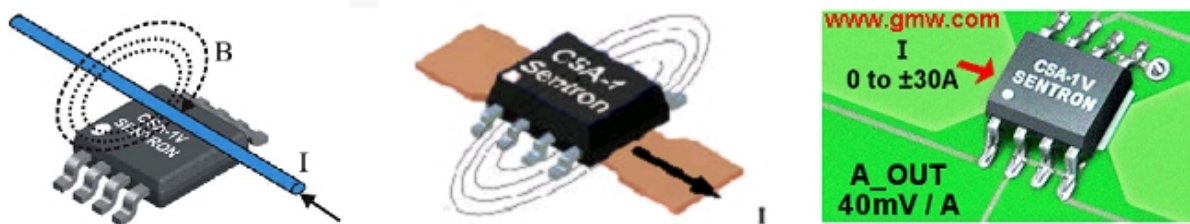


Fig. 63 How the CSA-1V Sentron works

According to the datasheet, when the current carrying conductor is about 0.1mm, 10A will produce 4V and the output terminal of the sensor chip. 5A would produce about 3.2 V.

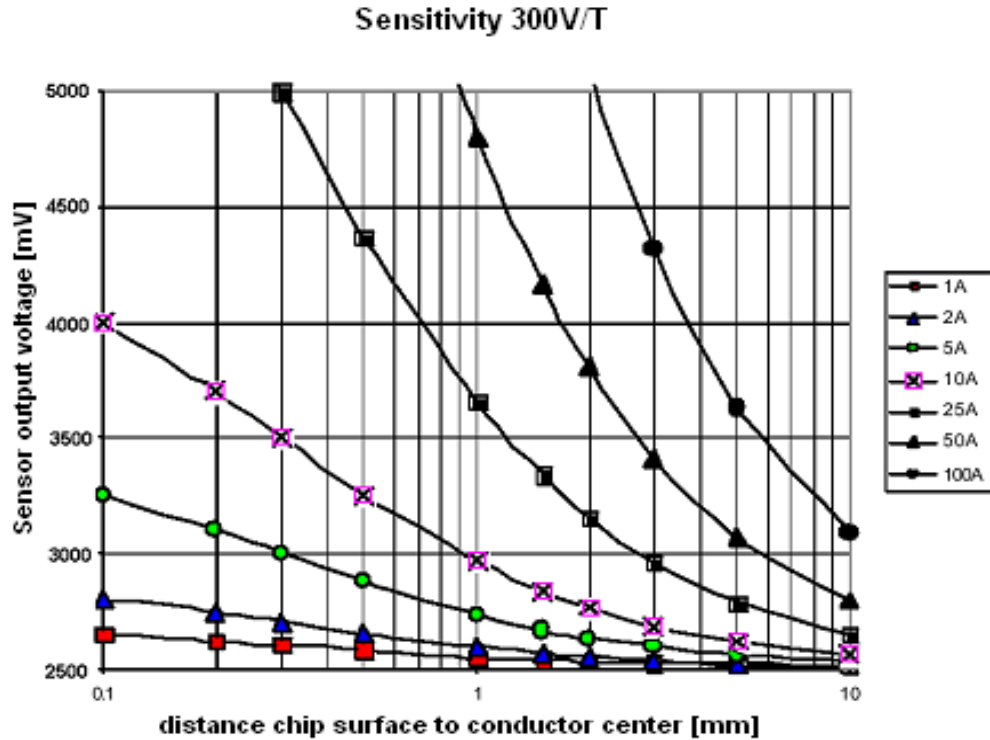


Fig. 64 Sensitivity graph

With the following connection, we can measure the current through the wire using the AN1 channel of 16F877. We apply the same logic and example code. The wire indicated with different color and different thickness (or AWG number) do not indicate the actual different in the conductor type on top of the sensor chip. We can use any wire, which can stand the maximum expected current flowing through, for the measurement. No special wire type is needed to place it on top (or bottom) of the sensor chip. However, you may want to use insulated wire to prevent an electric shock or electrocution by accidentally touching a bare wire carrying very high current.

There are two ways to get the output voltage for current measurement, if we get the output between pin#1 A_out and the ground, as connected in the drawing, the voltage range is [0, +5]V. However, if you use the pin#8 CO_out as internal reference, we can read a differential output voltage between pin #1 and pin#8, in which case the output range is [-2.5, +2.5]V.

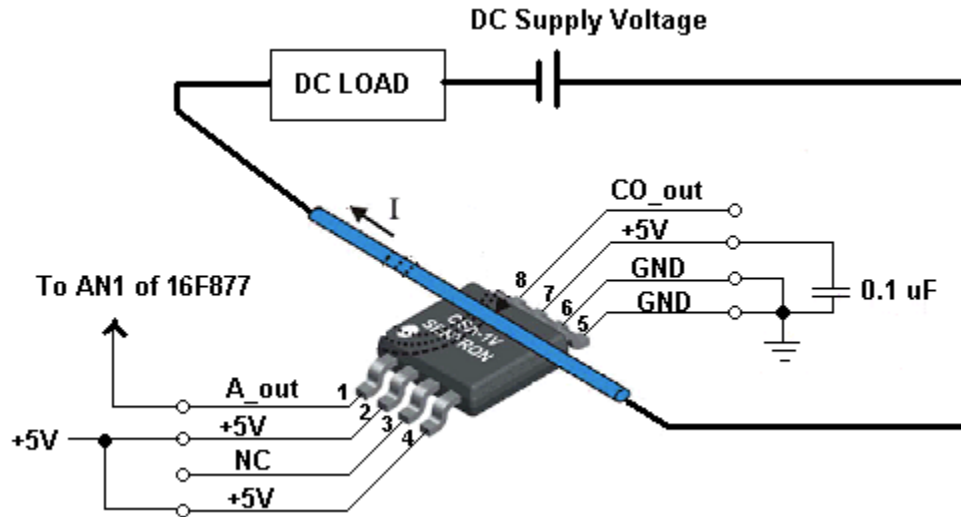


Fig. 65 CSA-1V Sentron connection to get output voltage

The next current sensing device is Maxim IC's MAX471/MAX472. MAX471 is a bidirectional, high-side current-sense amplifiers for portable PCs, telephones, and other systems where battery/DC power-line monitoring is critical. High-side power-line monitoring is especially useful in battery-powered systems, since it does not interfere with the ground paths of the battery chargers or monitors often found in "smart" batteries.

The MAX471 has an internal $35\text{m}\Omega$ current-sense resistor and measures battery currents up to $\pm 3\text{A}$. For applications requiring higher current or increased flexibility, the MAX472 functions with external sense and gain-setting resistors. Both devices have a current output that can be converted to a ground-referred voltage with a single resistor, allowing a wide range of battery voltages and currents.

An open-collector SIGN output indicates current-flow direction, so the user can monitor whether a battery is being charged or discharged. Both devices operate from 3V to 36V, draw less than $100\mu\text{A}$ over temperature, and include a $18\mu\text{A}$ max shutdown mode.

With the following connection, with $2\text{K}\Omega$ resistor at the output side, the current-to-voltage conversion ratio is 1V/A . Therefore, 1 V reading means 1 A current flow from the battery. The SIGN pin is to indicate the current flow direction. The Low level indicates that current flows from RS- to RS+.

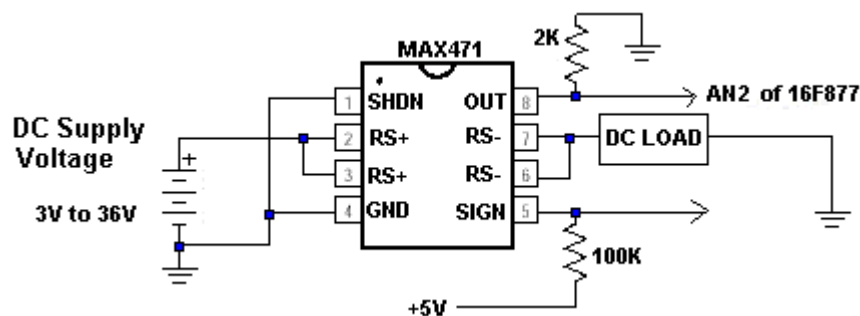


Fig. 66 Maxim IC's MAX471

In the connection diagram above, since the current flows, out of the DC supply voltage source, from RS+ to RD- to the LOAD to the ground, the SIGN output must be High. When we do not need the flow direction information, leave the pin open.

Another current sensor is Allegro MicroSystem's ACS750, a fully integrated current sensor. The Allegro ACS750 family of current sensors provides economical and precise solutions for current sensing in industrial, commercial, automotive, and communications systems. The device package allows for easy implementation by the customer. Typical applications include motor control, load detection and management, switched mode power supplies and over-current fault protection.

The sensor consists of a precision linear Hall IC optimized to an internal magnetic circuit to increase device sensitivity. The primary conductor used for current sensing (terminals 4 and 5) is designed for extremely low power loss. The power terminals are also electrically isolated from the sensor leads (pins 1 – 3). This allows the ACS750 family of sensors to be used in applications requiring electrical isolation without the use of opto-isolators or other costly isolation techniques.

As we see the current vs. output voltage curve, they are linearly related in the range of [0, +5]V for the current range of [-50, +50]A.

Allegro's ACS750 Current Sensor

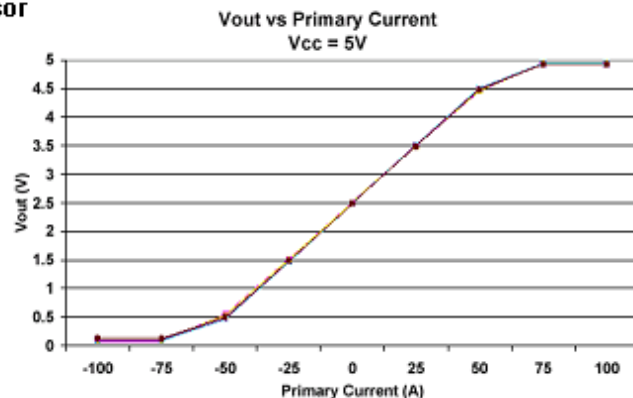
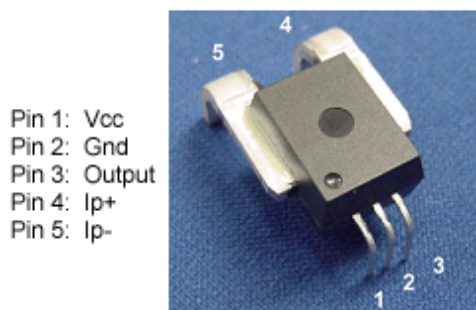


Fig. 67(a) Allegro MicroSystem's ACS750

Fig. 67(b) Graph of Volt vs. Primary Current

The following connection would read the current through the DC load in terms of voltage at the output pin #3, which is measured by the AN2 channel of the A/D conversion module. We use the same code we already examined for a wiper voltage measurement of a variable resistor.

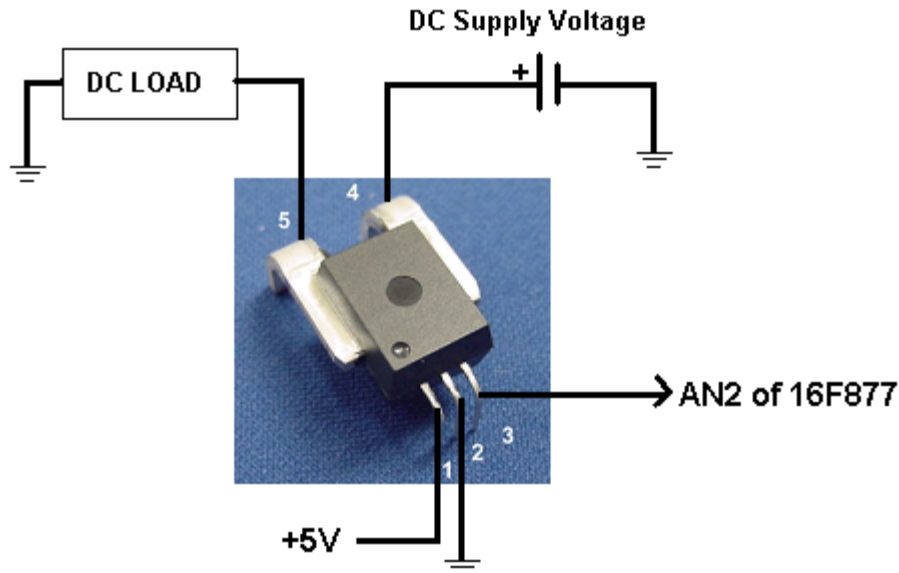


Fig. 68 ACS750 connection to read current through DC load

The last item we apply the same A/D conversion is a thermistor, or thermally sensitive resistor. The resistance value of the thermistor changes according to temperature. This part is used as a temperature sensor. There are three types of thermistors: negative temperature coefficient thermistor (NTC), positive temperature coefficient thermistor (PTC), and critical temperature resistor thermistor (CTR). NTC decreases its resistance value continuously as temperature rises. PTC increases its resistance value suddenly when temperature rises above a specific point. On the other hand, CTR decreases its resistance value when temperature rises above a certain point.

The thermistor we examine is a very small NTC thermistor, A170. Cool resistance is about $1.5\text{K}\Omega$ and hot resistance is about $25\ \Omega$. We may need some type of calibration to accurately convert the resistance of the thermistor to the ambient temperature. The following connection would get about 2.5V for normal temperature. As the temperature rises, the output voltage would also rise. Test the voltage output with the example code we studied for the wiper voltage of a variable resistor.

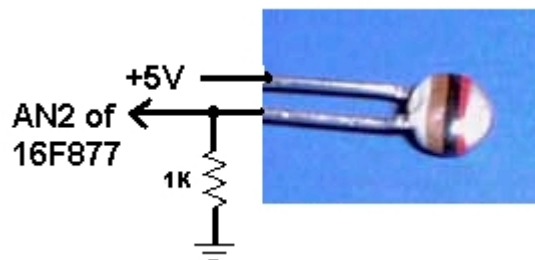


Fig. 69 NTC thermistor A170 connection to PIC 16F877